

1. For a bicycle model, the linearized equations are $M \ddot{q} + v C_1 \dot{q} + [g k_0 + v^2 k_2] q = f$, where $q = [\phi, \delta]^T$, $f = [T_\phi, T_\delta]^T$, $M = \begin{bmatrix} 80.8 & 2.3 \\ 2.3 & 0.3 \end{bmatrix}$, $g = 9.8$, $v = 2$, $C_1 = \begin{bmatrix} 0 & 33.9 \\ -0.9 & 1.7 \end{bmatrix}$, $k_0 = \begin{bmatrix} -81 & -2.6 \\ -2.6 & -0.8 \end{bmatrix}$, $k_2 = \begin{bmatrix} 0 & 76.6 \\ 0 & 2.7 \end{bmatrix}$.

(T_ϕ is lean torque, T_δ is steering torque)

a) Using $x = [\phi, \delta, \dot{\phi}, \dot{\delta}]^T$, $u = f$, write the state-space model $\dot{x} = Ax + Bu$. $A = ?$, $B = ?$

b) Is the system controllable using the input T_ϕ only ($T_\delta \equiv 0$, it is not there).

c) Is the system controllable using the input T_δ only ($T_\phi \equiv 0$, it is not there).

a) $x_1 = \phi, x_2 = \delta, x_3 = \dot{\phi}, x_4 = \dot{\delta}$
 $\Rightarrow \dot{x}_1 = x_3, \dot{x}_2 = x_4$

$$M \ddot{q} + v C_1 \dot{q} + [g k_0 + v^2 k_2] q = f$$

$$\Rightarrow M \begin{bmatrix} \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} + v C_1 \begin{bmatrix} x_3 \\ x_4 \end{bmatrix} + [g k_0 + v^2 k_2] \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = u$$

$$\Rightarrow M \begin{bmatrix} \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} = -[g k_0 + v^2 k_2] \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} - v C_1 \begin{bmatrix} x_3 \\ x_4 \end{bmatrix} + u$$

$$\Rightarrow \begin{bmatrix} \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} = -M^{-1} [g k_0 + v^2 k_2] \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} - M^{-1} v C_1 \begin{bmatrix} x_3 \\ x_4 \end{bmatrix} + M^{-1} u$$

$$\therefore \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ \dots & \dots & \dots & \dots \\ -M^{-1}x & 1 & -M^{-1}x \\ [gk_0 + v^2 k_2] & v C_1 \end{bmatrix}}_A \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \underbrace{\begin{bmatrix} 0 & 0 \\ 0 & 0 \\ \dots & \dots \\ M^{-1} \end{bmatrix}}_B u$$

Numerical values are, $M = \begin{bmatrix} 80.8 & 2.3 \\ 2.3 & 0.3 \end{bmatrix}$, $g = 9.8$,
 $v = 2$, $C_1 = \begin{bmatrix} 0 & 33.9 \\ -0.9 & 1.7 \end{bmatrix}$, $k_0 = \begin{bmatrix} -81 & -2.6 \\ -2.6 & -0.8 \end{bmatrix}$, $k_2 = \begin{bmatrix} 0 & 76.6 \\ 0 & 2.7 \end{bmatrix}$.

$$M^{-1} = \frac{1}{\begin{matrix} 80.8 \times 0.3 & - (2.3)^2 \\ 24.24 & 5.29 \end{matrix}} \begin{bmatrix} 0.3 & -2.3 \\ -2.3 & 80.8 \end{bmatrix}$$

$$= 0.053 \begin{bmatrix} \quad \quad \quad \\ \quad \quad \quad \end{bmatrix} = \begin{bmatrix} 0.016 & -0.122 \\ -0.122 & 4.28 \end{bmatrix}$$

$$M^{-1} v C_1 = 2 \begin{bmatrix} 0.016 & -0.122 \\ -0.122 & 4.28 \end{bmatrix} \begin{bmatrix} 0 & 33.9 \\ -0.9 & 1.7 \end{bmatrix}$$

$$= \begin{bmatrix} 2(0.122 \times 0.9) & 2(0.016 \times 33.9 - 0.122 \times 1.7) \\ -2 \times 4.28 \times 0.9 & -2(33.9 \times 0.122 - 4.28 \times 1.7) \end{bmatrix}$$

$$\begin{matrix} 0.542 & 0.207 \\ 4.14 & 7.28 \end{matrix}$$

$$= \begin{bmatrix} 0.22 & 0.67 \\ -7.7 & 6.28 \end{bmatrix}$$

$$M^{-1} [g K_0 + v^2 K_2] = M^{-1} \left(\begin{array}{cc} -81 \times 9.8 & -2.6 \times 9.8 \\ -2.6 \times 9.8 & -0.8 \times 9.8 \end{array} \right) + \left(\begin{array}{cc} 0 & 4 \times 76.6 \\ 0 & 4 \times 2.7 \end{array} \right)$$

$$= \begin{bmatrix} 0.016 & -0.122 \\ -0.122 & 4.28 \end{bmatrix} \begin{bmatrix} -793.8 & 280.92 \\ -25.48 & 2.96 \end{bmatrix}$$

$$= \begin{bmatrix} -0.016 \times 793.8 + 0.122 \times 25.48 & 280.92 \times 0.016 - 0.122 \times 2.96 \\ 0.122 \times 793.8 - 4.28 \times 25.48 & -280.92 \times 0.122 + 4.28 \times 2.96 \end{bmatrix}$$

$$= \begin{bmatrix} -9.58 & 4.13 \\ -12.25 & -21.6 \end{bmatrix}$$

$$A = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 9.58 & -4.13 & -0.22 & -0.67 \\ 12.25 & 21.6 & 7.7 & -6.28 \end{bmatrix}$$

$$B = \left[\begin{array}{c|c} b_1 & b_2 \\ \hline 0 & 0 \\ 0 & 0 \\ 0.016 & -0.122 \\ -0.122 & 4.28 \end{array} \right]$$

b) & c) check
 rank $[b_i \ A b_i \ A^2 b_i \ A^3 b_i] = 4$
 $i=1$ for b)
 $i=2$ for c)

2. A and B are square matrices.

a) Show that, if $AB=BA$, $e^{A+B} = e^A \cdot e^B$.

b) By counterexample, choose A and B such that $AB \neq BA$ and check that $e^{A+B} \neq e^A \cdot e^B$

$$a) e^{A+B} = I + (A+B) + \frac{(A+B)^2}{2!} + \frac{(A+B)^3}{3!} + \dots$$

$$e^A = I + A + \frac{A^2}{2!} + \frac{A^3}{3!} + \dots$$

$$e^B = I + B + \frac{B^2}{2!} + \frac{B^3}{3!} + \dots$$

$$e^A \cdot e^B = I + (A+B) + \left(\frac{A^2}{2!} + \frac{B^2}{2!} + AB \right) + \dots$$

$$\text{Now, } (A+B)^2 = (A+B)(A+B) \\ = A^2 + AB + BA + B^2$$

If $AB=BA$, then $(A+B)^2 = A^2 + 2AB + B^2$
and $e^{A+B} = e^A \cdot e^B$

(similarly for higher powers of $(A+B)$)

$$b) A = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}, \quad B = \begin{bmatrix} 0 & -1 \\ 0 & 1 \end{bmatrix}$$

$$AB = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}, \quad BA = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}$$

Here $AB \neq BA$.

$$A+B = \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix}$$

$$e^A = I + A \quad \left[\because A^2 = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \right]$$
$$= \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 0 & 0 \end{bmatrix}$$

$$e^B = I + B + \frac{B^2}{2!} + \dots \quad \left[B^2 = \begin{bmatrix} 0 & -1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 0 & -1 \\ 0 & 1 \end{bmatrix} \right]$$
$$= I + B \left(1 + \frac{1}{2!} + \frac{1}{3!} + \dots \right) = \begin{bmatrix} 0 & -1 \\ 0 & 1 \end{bmatrix} = B$$

$$= I + B(e-1)$$

$$= \begin{bmatrix} 1 & 1-e \\ 0 & e \end{bmatrix}$$

$$e^A \cdot e^B = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 1 & 1-e \\ 0 & e \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ 0 & e \end{bmatrix}$$

$$e^{A+B} = I + (A+B)(e-1) \quad \left[\because (A+B)^2 = \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix} \right]$$
$$= \begin{bmatrix} 1 & 0 \\ 0 & e \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix}$$

3. A disastrous way to control an unstable system is to cancel the unstable pole with a zero. Consider a transfer function from input to output as $H(s) = \frac{1}{s^2 - \Omega^2}$

a) Is this unstable?

Consider a controller $G(s) = \frac{s - \bar{\Omega}}{s + \bar{\Omega}}$.

Clearly, $G(s)H(s) = \frac{1}{s(s + \Omega)}$ when $\Omega = \bar{\Omega}$.

When $\Omega \neq \bar{\Omega}$, the transfer function $G(s)H(s)$ can be shown to be in the form $\dot{x} = Ax + Bu$,

where $A = \begin{bmatrix} 0 & 1 & 0 \\ -\Omega^2 & 0 & -1 \\ 0 & 0 & 0 \end{bmatrix}$, $B = \begin{bmatrix} 0 \\ 1 \\ \bar{\Omega} \end{bmatrix}$.

b) Find the transformation T that diagonalizes A . $AT = TD$

c) Express the equation in z -coordinates, where $z = T^{-1}x$. ($z = [z_1, z_2, z_3]'$)

d) Analyze the z -equations in the limit $\bar{\Omega} \rightarrow \Omega$. Can they be controlled?

a) Unstable as pole = +1 in RHP

b) eigenvalues of A ,

$$sI - A = \begin{bmatrix} s & -1 & 0 \\ -\Omega^2 & s & 1 \\ 0 & 0 & s \end{bmatrix}$$

$$\det(sI - A) = s(s^2 + \Omega^2)$$

\therefore eigenvalues are $\Omega, 0, -\Omega$

eigenvectors:

$$\underline{s = \Omega}, \quad sI - A = \begin{bmatrix} \Omega & -1 & 0 \\ -\Omega^2 & \Omega & -1 \\ 0 & 0 & \Omega \end{bmatrix}$$

$$\sim \begin{bmatrix} \Omega & -1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & \Omega \end{bmatrix} \sim \begin{bmatrix} \Omega & -1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{bmatrix}$$

\therefore eigenvector can be $\begin{bmatrix} 1 \\ \Omega \\ 0 \end{bmatrix}$

$$\underline{s = 0}, \quad sI - A = \begin{bmatrix} 0 & -1 & 0 \\ -\Omega^2 & 0 & -1 \\ 0 & 0 & 0 \end{bmatrix}$$

\therefore eigenvector can be $\begin{bmatrix} 1 \\ 0 \\ \Omega^2 \end{bmatrix}$

$$\underline{s = -\Omega}, \quad sI - A = \begin{bmatrix} -\Omega & -1 & 0 \\ -\Omega^2 & -\Omega & 1 \\ 0 & 0 & -\Omega \end{bmatrix}$$

$$\sim \begin{bmatrix} -\Omega & -1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & -\Omega \end{bmatrix} \sim \begin{bmatrix} \Omega & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{bmatrix}$$

\therefore eigenvector can be $\begin{bmatrix} 1 \\ -\Omega \\ 0 \end{bmatrix}$

The transformation matrix, $T = \begin{bmatrix} 1 & 1 & 1 \\ -\Omega & 0 & -\Omega \\ 0 & \Omega^2 & 0 \end{bmatrix}$

c) $\dot{z} = T^{-1} \dot{x}$
 $= \underbrace{T^{-1} A T}_{\rightarrow} z + T^{-1} B u$

$$T^{-1} = \frac{\text{adj}(T)}{\det(T)} \quad \begin{bmatrix} \Omega & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & -\Omega \end{bmatrix}$$

$$\det(T) = -\Omega^2 \begin{vmatrix} 1 & 1 \\ \Omega & -\Omega \end{vmatrix} = 2\Omega^3$$

$$\text{adj}(T) = \begin{bmatrix} \Omega^3 & 0 & \Omega^3 \\ \Omega^2 & 0 & -\Omega^2 \\ -\Omega & 2\Omega & -\Omega \end{bmatrix}^T = \begin{bmatrix} \Omega^3 & \Omega^2 & -\Omega \\ 0 & 0 & 2\Omega \\ \Omega^3 & -\Omega^2 & -\Omega \end{bmatrix}$$

$$\therefore T^{-1} B = \begin{bmatrix} \frac{1}{2} & \frac{1}{2\Omega} & \frac{1}{2\Omega^2} \\ 0 & 0 & \frac{1}{\Omega^2} \\ \frac{1}{2} & \frac{1}{2\Omega} & \frac{1}{\Omega^2} \end{bmatrix} \begin{bmatrix} \Omega \\ 1 \\ 0 \end{bmatrix} = \begin{bmatrix} \frac{1}{2\Omega^2}(\Omega - \Omega) \\ \frac{1}{\Omega^2} \\ \frac{1}{2\Omega^2}(\Omega + \Omega) \end{bmatrix}$$

as $\bar{\Omega} \rightarrow \Omega$,

$$\dot{z}_1 = \Omega z_1 + \frac{1}{2\Omega^2} (\Omega - \bar{\Omega}) u$$

$$\dot{z}_2 = \frac{\bar{\Omega}}{\Omega^2} u$$

$$\dot{z}_3 = -\Omega z_3 - \frac{1}{2\Omega^2} (\Omega + \bar{\Omega}) u$$

$\rightarrow 0$
pole-zero
cancellation

z_1 - equation is unstable and there is no input term