

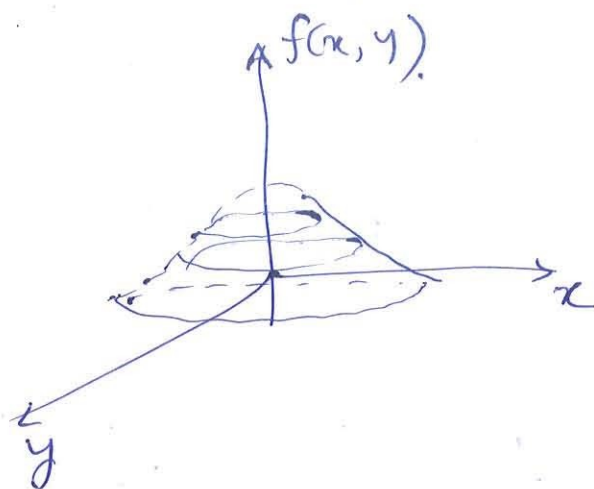
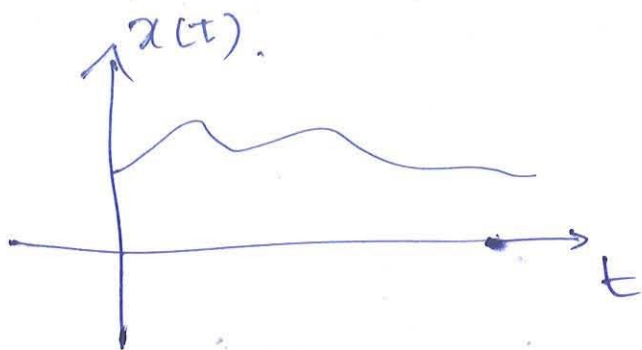
Signals & systems. (Week 1)

①

- What are signals ?

Traffic signal [green, ~~blue~~, yellow, red].
Sound, Temperature, Pressure, velocity, acceleration, etc.

- A measurement of physical quantities ~~as~~ represented as their variation with time, space, or any other independent direction.



- Most applications requires us to deal with or treat them as a system that operates on signals and gives out the desired signal.

- Broadly can be classified as analysing signals, extracting, ~~or~~ shaping signals.

eg:- Predict weather conditions based on past observations of pressure, temperature, wind speed etc.

- Seismic activity monitoring

- Voltage, currents, load angles on power grid. Predicting worst case scenarios.

- get estimate of some indirectly available signals. Medical applications (ultrasound, ECG, MRI, etc)

- Shape signals to get desired behaviour. Noise cancelling headphones, hearing aids, etc.

This course is building block which will teach to ~~analyse & design~~ how to effectively deal with signals & system.

Signal & its representation

(2)

Signal is a function of "possibly" several independent variables.

$$f: \mathbb{R}^m \rightarrow \mathbb{R}^n$$

$$\begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_m \end{bmatrix} \mapsto \begin{bmatrix} f_1(x_1, x_2, \dots, x_m) \\ f_2(x_1, x_2, \dots, x_m) \\ \vdots \\ f_n(x_1, x_2, \dots, x_m) \end{bmatrix}$$

eg: Video signal is a signal in

(x, y) and t where (x, y) is pixel location and t is frame number

$\begin{bmatrix} V_C(x, y, t) \\ V_I(x, y, t) \end{bmatrix}$ gives the color & intensity of pixel (x, y) at frame t .

$$V: \mathbb{R}^3 \rightarrow \mathbb{R}^2$$

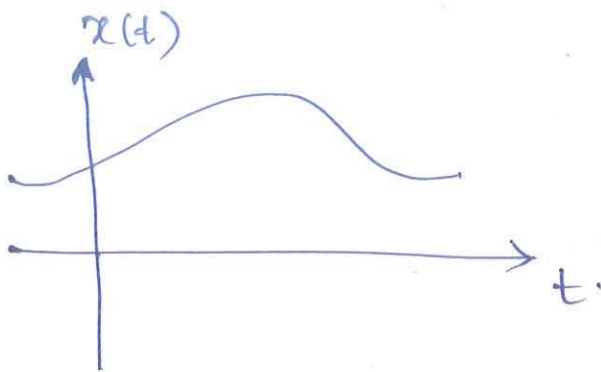
In this course we will deal with signals which are functions of "time" variable.

notation: $f(t)$, $v(t)$, ω , etc.

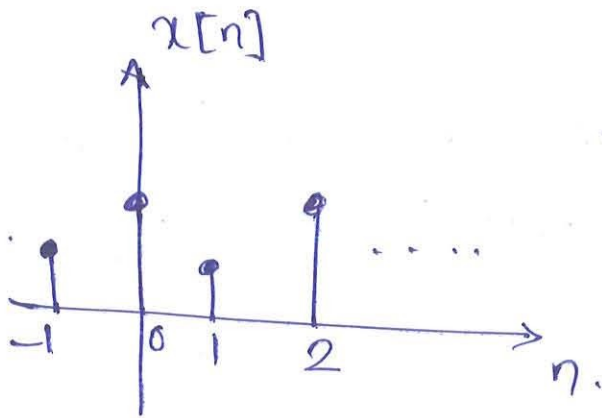
Signals can be classified in two

Categories

- (i) Continuous time Signals. [notation: $x(t), u(t), y(t)$]
- (ii) Discrete time Signals. [notation: $x[n], u[n], i[n]$]



Continuous time signals are the ones where the value of signal is specified at every $t \in \mathbb{R}$
eg: Analog ~~continuous~~ Signals.



Discrete time signals are the ones where value of signal is specified only at some discrete "samples" of time.
eg: Digital signals can be represented in this manner.

Also those quantities that cannot be monitored continuously gives rise to discrete ^{time} signals.
for eg: - Stock market indices, Blood glucose levels, etc.

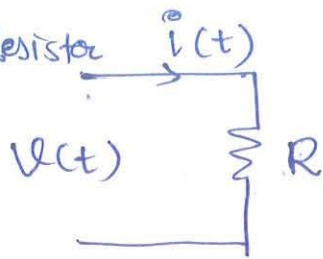
- Signal Energy and Power.

[May not be always related to physical energy/
power notions. Must be treated in

abstract manner.] But these notions are
indeed "inspired" from the physical notions.
(or generalised)

Example: Instantaneous Power supplied to Resistor

$$p(t) = \frac{v^2(t)}{R}$$



Then $\int_{t_1}^{t_2} \frac{v^2(t)}{R} dt$ ~~is~~ ^{is} be the total

energy expended in $t \in [t_1, t_2]$

and $\frac{1}{t_2 - t_1} \int_{t_1}^{t_2} \frac{v^2(t)}{R} dt$ is average

power in this interval

Let $R=1$. Then,

We can say that a signal $v(t)$ has
total energy content in $[t_1, t_2]$ to be

$$\int_{t_1}^{t_2} v^2(t) dt.$$

~~total~~ average power $\frac{1}{t_2 - t_1} \int_{t_1}^{t_2} v^2(t) dt$

Generalizing for any signal $x(t)$.

Total energy on interval $[t_1, t_2]$:

$$\int_{t_1}^{t_2} |x(t)|^2 dt$$

$|x(t)|$ is the absolute value of complex no. at time t .
(In case of complex signal)

Time Averaged ~~time~~ power

$$\frac{1}{t_2 - t_1} \int_{t_1}^{t_2} |x(t)|^2 dt$$

Analogously for Discrete time signals.

Energy on samples between $[n_1, n_2]$

$$\sum_{n=n_1}^{n_2} |x[n]|^2$$

Average power

$$\frac{1}{n_2 - n_1 + 1} \sum_{n=n_1}^{n_2} |x[n]|^2$$

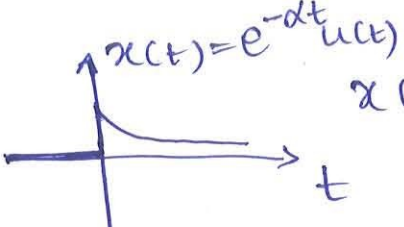
~~Plot~~

We may be sometimes interested in Energy over infinite interval the so-called full line ~~intervals~~ intervals $(-\infty, \infty)$

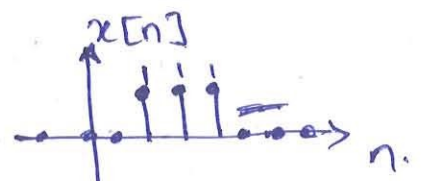
Continuous time:
$$E_{\infty} := \lim_{T \rightarrow \infty} \int_{-T}^T |x(t)|^2 dt = \int_{-\infty}^{\infty} |x(t)|^2 dt$$

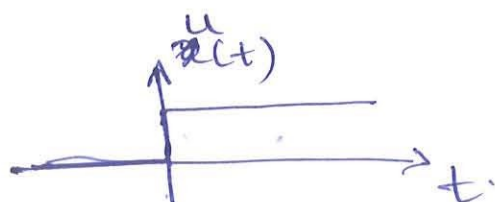
[Integration is finite ~~the~~ finite energy signal.

Discrete time:
$$E_{\infty} := \lim_{N \rightarrow \infty} \sum_{n=-N}^N |x[n]|^2 \quad \left[\begin{array}{l} \text{Summation finite} \\ \text{finite energy signal} \end{array} \right]$$

Examples: 
$$x(t) = \begin{cases} 0, & t < 0 \\ e^{-\alpha t}, & t \geq 0 \\ \alpha > 0 \end{cases}$$

is a finite energy signal.

or  is also a finite energy signal.

but 
$$x(t) = \begin{cases} 0 & t < 0 \\ 1 & t \geq 0 \end{cases}$$

is infinite energy signal.

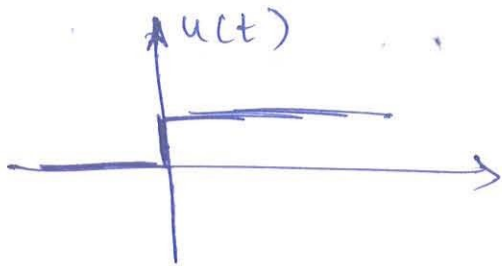
Time averaged power over infinite interval.

Continuous time:

$$P_{av} = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T |x(t)|^2 dt$$

Discrete time:

$$P_{av} = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N |x[n]|^2$$



is a finite average power signal.

[But infinite energy]

Verify for $(-\infty, \infty)$ interval.

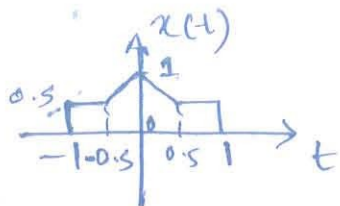
- Finite energy signal gives zero average power.

- Finite average power signal leads to infinite energy.

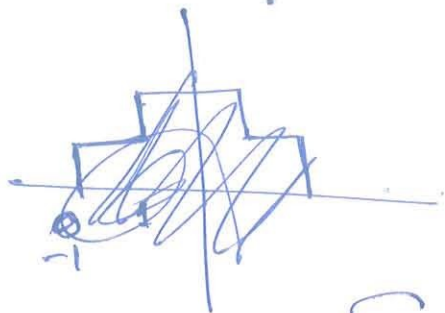
What about $x(t) = t$?

Signal Transformations.

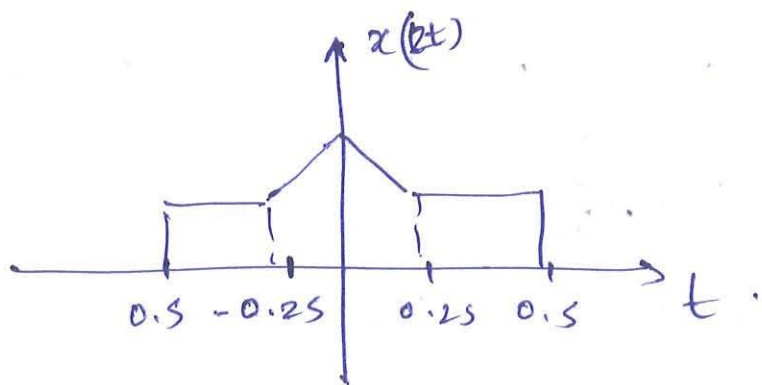
$x(t)$



$$x(t) = \begin{cases} 0, & t < -1 \\ 0.5, & -1 \leq t < -0.5 \\ t+1, & -0.5 \leq t \leq 0 \\ t+1, & 0 \leq t < 0.5 \\ 0.5, & 0.5 \leq t \leq 1 \\ 0, & t > 1 \end{cases}$$



$$x(2t) = \begin{cases} 0, & 2t < -1 \Rightarrow t < -0.5 \\ 0.5, & -1 \leq 2t \leq -0.5 \Rightarrow -0.5 \leq t \leq -0.25 \\ 2t+1, & -0.5 \leq 2t \leq 0 \Rightarrow -0.25 \leq t \leq 0 \\ 2t-1, & 0 \leq 2t < 0.5 \Rightarrow 0 \leq t < 0.25 \\ 0.5, & 0.5 \leq 2t \leq 1 \Rightarrow 0.25 \leq t \leq 0.5 \\ 0, & 2t > 1 \Rightarrow t > 0.5 \end{cases}$$



Similarly,
Calculate $x\left(\frac{t}{2}\right)$, $x(-t)$, $x(t-1)$, $x(t+1)$
and so on.

Periodic Signals.

Most often encountered.

- Harmonic oscillators, ~~etc.~~
- Seasonal cycles, predator prey.
- Clocks - Extensively used in Fourier analysis (later).

Designing a stable clock was a big problem in past

A signal $x(t)$ is a periodic signal if

there is a positive no. $T > 0$ s.t.

$$x(t+T) = x(t)$$

if $x(t+T) = x(t)$

then $x(t+mT) = x(t)$ for any

integer $m \in \mathbb{Z}$

\mathbb{Z} is a set of integers

which means any mT is time period of $x(t)$ if T is a time period.

- The least ^{positive} number $T_0 > 0$ for which,

$x(t+T_0) = x(t)$ is known as the fundamental period of $x(t)$.

eg: $\sin(2\pi t) = \sin(2\pi(t+m))$ for any $m \in \mathbb{Z}$

But ~~1~~ $T_0 = 1$ is least no. s.t

$$\sin(2\pi(t+T_0)) = \sin 2\pi t$$

Thus 1 is the fundamental time period of $\sin 2\pi t$

In Discrete time signals one needs to be careful.

A discrete time signal $x[n]$ is periodic if there exists a positive "integer" $N > 0$ s.t.

$$x[n+N] = x[n].$$

Just like in continuous case the least positive

"integer" N_0 s.t.

$$x[n+N_0] = x[n] \text{ is fundamental}$$

period.

eg: $\cos\left(\frac{2\pi n}{3}\right)$ is periodic with

period. $N_0 = 3$

$$\begin{aligned} \cos\left(\frac{2\pi}{3}(n+3)\right) &= \cos\left(\frac{2\pi n}{3} + 2\pi\right) \\ &= \cos\left(\frac{2\pi n}{3}\right) \end{aligned}$$

But $\cos\left(\frac{n}{2}\right)$ is not periodic

because there is no "integer"

$$\text{s.t. } \cos\left(\frac{n+N}{2}\right) = \cos\left(\frac{n}{2}\right)$$

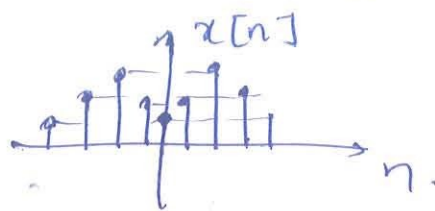
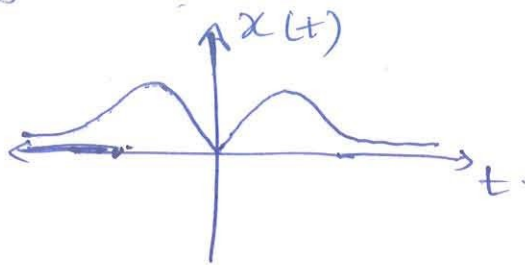
- Even and Odd Signals.

Even signal. $x(t)$ Satisfies

$$x(t) = x(-t) \text{ (continuous time)}$$

$$x[n] = x[-n] \text{ (discrete time)}$$

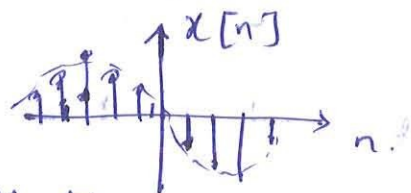
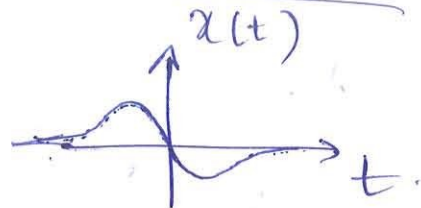
Symmetric about $t=0$ (or $n=0$)



Odd signals.

$$x(t) = -x(-t)$$

$$x[n] = -x[-n]$$



for odd signals notice that

$$x(0) = -x(0) \Rightarrow x(0) = 0$$

and $x[0] = 0$.

Necessarily Zero at the origin

$$x(t) = \frac{x(t) + x(-t)}{2} + \frac{x(t) - x(-t)}{2}$$

$$\underbrace{\hspace{10em}}$$

Ev($x(t)$)

Even part

$$\underbrace{\hspace{10em}}$$

Od($x(t)$)

odd part.

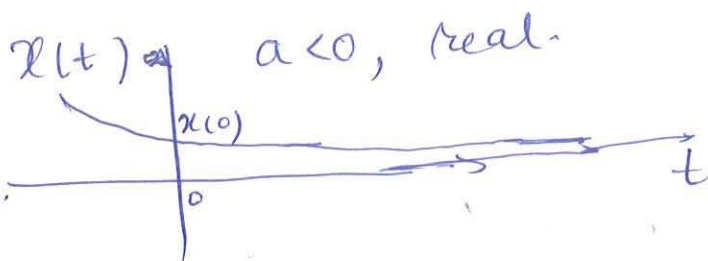
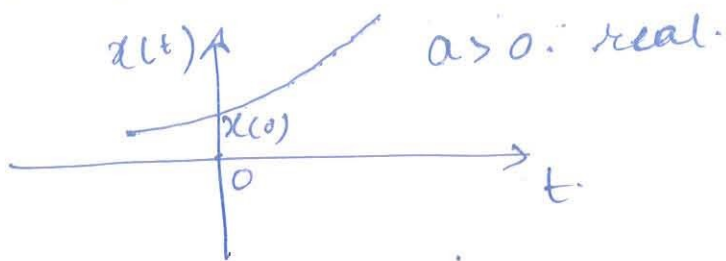
$$\frac{x(-t) - x(+t)}{2} = -\frac{(x(t) - x(-t))}{2}$$

Since $\frac{x(-t) + x(t)}{2}$

Commonly Encountered signals.

(i) Exponential signals.

$$x(t) = x(0) e^{at} \quad \left["a" \text{ can be complex.} \right]$$



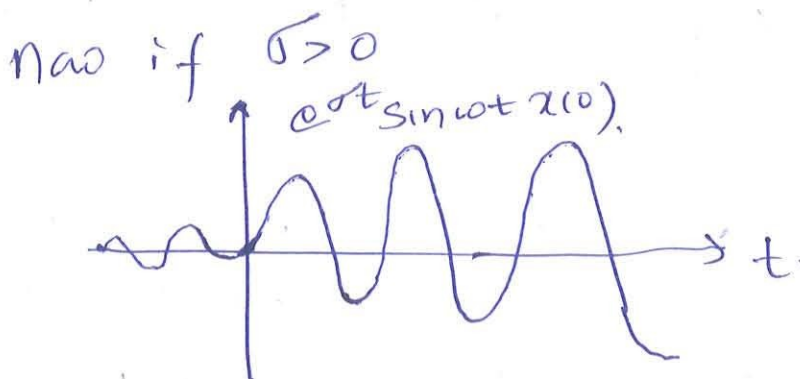
a complex.

then

$$x(t) = e^{(\sigma + j\omega)t} x(0)$$
$$= e^{\sigma t} (\cos \omega t + j \sin \omega t) x(0)$$

[Euler's identity]

$$e^{j\omega t} = \cos \omega t + j \sin \omega t$$



Similar plot
 $e^{\sigma t} \cos \omega t$

$$e^{j\omega_0 t} = e^{j\omega_0 (t + \frac{2\pi}{\omega_0})}$$

Thus, $T_0 = \frac{2\pi}{\omega_0}$ is fundamental period.

In general, $\omega_0 T = 2\pi m$, $m \in \mathbb{Z}$ is a time period for $e^{j\omega_0 t}$, but least

T is $T_0 = \frac{2\pi}{\omega_0}$ fundamental period.

Sine & Cosine with Phase.

$$x(t) = A \cos(\omega_0 t + \phi)$$

$T_0 = \frac{2\pi}{\omega_0}$ is fundamental period.

DISCRETE TIME

$$x[n] = e^{j\omega_0 n}$$

$$e^{j\omega_0 (n+N)} = e^{j\omega_0 n}$$

$$\Rightarrow e^{j\omega_0 N} = 1$$

Q will be satisfied if $\omega_0 N = 2\pi m$, $m \in \mathbb{Z}$

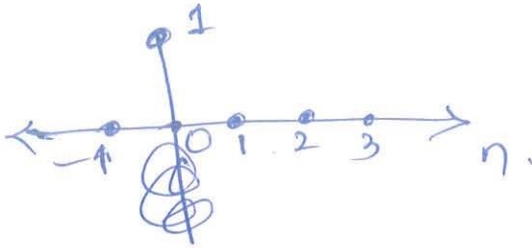
Thus, if $N = \frac{2\pi m}{\omega_0}$ is an integer

only, then N is the period of signal $x[n]$.

If integer, $N_0 = \frac{2\pi}{\omega_0}$ is fundamental period.

- Unit impulse. [Discrete time].

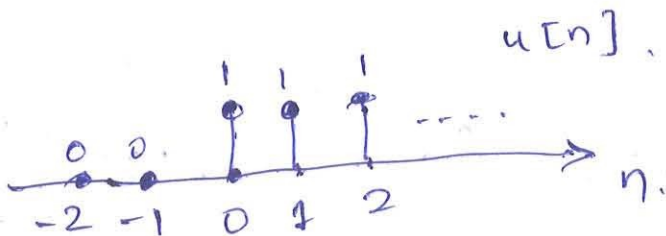
$$\delta[n] = \begin{cases} 0, & n \neq 0 \\ 1, & n = 0 \end{cases}$$



plot $\delta[n-1]$
 $\delta[n+2]$?

- Unit step.

$$u[n] = \begin{cases} 0, & n < 0 \\ 1, & n \geq 0 \end{cases}$$



For any sequence or signal $x[n]$,
Pointwise multiplication.
 $x[n] \delta[n] = x[0] \delta[n]$.

$$x[n] \delta[n-2] = x[2] \delta[n-2]$$

Pointwise multiplication value of $x[n]$ at $n=2$ signal

$\delta[n]$ samples $x[n]$ at $n=0$.
 $\delta[n-2]$ samples $x[n]$ at $n=2$.

note:

$$\delta[n] = u[n] - u[n-1]$$

$$u[n] = \sum_{m=-\infty}^n \delta[m]$$

Cumulative. Sum of values of $\delta[m]$ at n .

$$u[n] = \sum_{k=-\infty}^0 \delta[n-k] = \sum_{k=0}^{\infty} \delta[n-k]$$

Continuous time unit step

$$u(t) = \begin{cases} 0, & t < 0 \\ 1, & t > 0 \end{cases}$$

undefined at $t=0$
(by convention?)

Dirac-delta (Ideal Impulse)

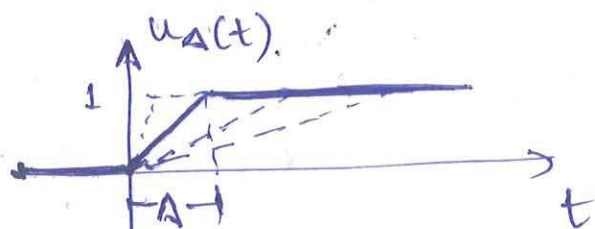
$$\boxed{\frac{d u(t)}{dt} = \delta(t)} \rightarrow \text{Strange!}$$

notice in discrete time case.
 $\delta[n] = u[n] - u[n-1]$

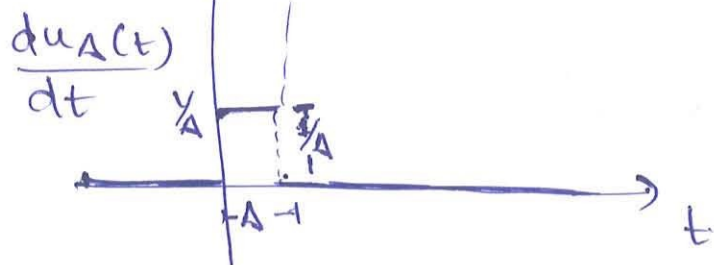
[derivative is to be understood in limiting sense]

[more precise defn. can be done by using "theory of distributions" (Schwarz, 1951)]

Consider an approximate unit step. $u_{\Delta}(t)$.



$$u_{\Delta}(t) = \begin{cases} 0 & t < 0 \\ \left(\frac{1}{\Delta}\right)t & 0 \leq t \leq \Delta \\ 1 & t > \Delta \end{cases}$$

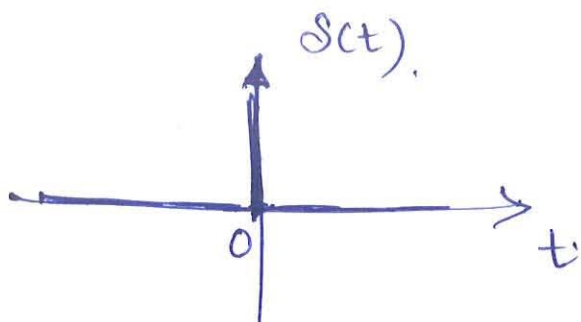


$$\frac{d u_{\Delta}(t)}{dt} = \begin{cases} 0, & t < 0 \\ \frac{1}{\Delta}, & 0 < t \leq \Delta \\ 0, & t > \Delta \end{cases}$$

Then $\frac{du(t)}{dt} = \lim_{\Delta \rightarrow 0} \frac{du_{\Delta}(t)}{dt} = \delta(t)$.

It is in this limiting sense that the derivating $\frac{du(t)}{dt}$ is to be understood.


We represent $\delta(t) = \begin{cases} \text{infinity} & , t = 0 \\ 0 & , t \neq 0 \end{cases}$



The ~~magnitude~~ ^{strength} of $\delta(t)$ is measured by ~~the~~ ^{its} area

Note that we can approximate $\delta(t)$ by $\delta_{\Delta}(t)$ which is $\frac{du_{\Delta}(t)}{dt}$ (see previous page)

Then $\int_{-\infty}^t \delta(t) dt = u(t)$ for $t > 0$. because $\int_{-\infty}^t \delta_{\Delta}(t) dt = 1$ for $t \neq \Delta > 0$ and as $\Delta \rightarrow 0$.

We could have very well use function  to approximate unit step. and arrived at same limiting case.

$\int_{-\infty}^t \delta_{\Delta}(t) dt \rightarrow \int_{-\infty}^t \delta(t) dt$ and $\int_{-\infty}^t \delta_{\Delta}(t) dt = 1$ so is $\int_{-\infty}^t \delta(t) dt$

Thus $\int_{-\infty}^t \delta(t) = u(t)$

and $\int_0^{\infty} \delta(t-\sigma) d\sigma = u(t)$.

properties of $\delta(t)$

$$\delta(-t) = \delta(t).$$

In fact $\delta(t)$ is sometimes defined as an object which acts on signal $x(t)$ and gives its value at particular instance. for example.

$$x(0) = \int_{-\infty}^{\infty} x(t) \delta(t) dt$$

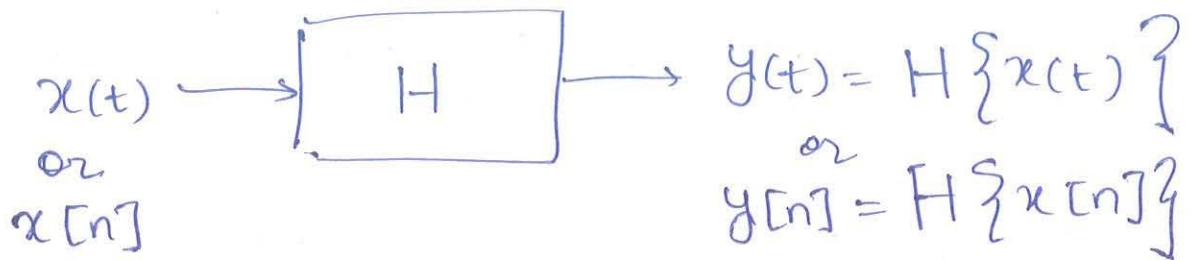
$$\text{or } x(x) = \int_{-\infty}^{\infty} x(t) \delta(t-x) dt$$

This property we will derive and use later to represent any signal $x(t)$ in terms of continuum of Dirac-delta impulses.

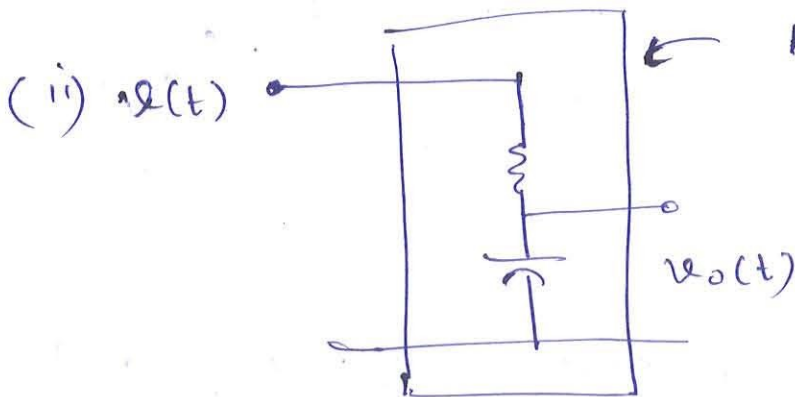
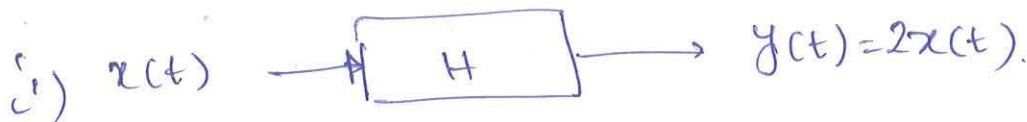
Systems.

— Systems. operate on signals and gives out another signals.

One can treat a system as a blackbox or a function H



Ex:



A slightly more complicated system.

— A ~~system~~ very useful class of systems that we encounter frequently are the systems that possess ~~the~~ following properties

- (i) linearity
- (ii) Time invariance.

- Linear system. [A similar defn holds for discrete-time systems]

A system H is said to be linear if it satisfies:

(i) for any signals $x_1(t)$ and $x_2(t)$

Suppose $y_1(t) = H\{x_1(t)\}$ and

$$y_2(t) = H\{x_2(t)\}.$$

then $y(t) = H\{x_1(t) + x_2(t)\}$.

$$= H\{x_1(t)\} + H\{x_2(t)\}$$

$$= y_1(t) + y_2(t).$$

holds.

[additivity/
Superposition]

(ii) for any scalar $\alpha \in \mathbb{C}$, and any signal $x(t)$

Suppose $y(t) = H\{x(t)\}$.

then $\bar{y}(t) = H\{\alpha x(t)\}$

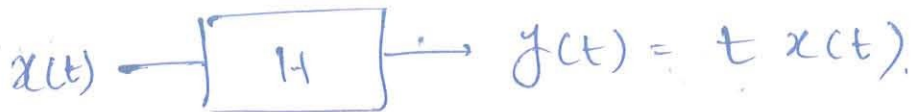
$$= \alpha H\{x(t)\} = \alpha y(t).$$

holds.

[Homogeneity property]

Note in particular ~~for~~ for $x(t) \equiv 0$, $y(t) \equiv 0$
(choose $\alpha = 0$ and apply homogeneity)

eg: $y(t) = \cancel{H\{x(t)\}} H\{x(t)\} = t x(t)$



Consider any $x_1(t)$ and $x_2(t)$

s.t. $y_1(t) = t x_1(t)$

$y_2(t) = t x_2(t)$

then indeed, for $x(t) = x_1(t) + x_2(t)$

$y(t) = H\{x_1(t) + x_2(t)\}$

$= \cancel{H\{t(x_1(t) + x_2(t))\}}$

$= t y_1(t) + y_2(t)$

Similarly for any α and $x(t)$

if $y(t) = H\{x(t)\} = t x(t)$

then $y(t) = H\{\alpha x(t)\}$

$= \alpha t x(t)$

$= \alpha t x(t) = \alpha y(t)$

hence, $y(t) = t x(t)$ is a linear system

Example 2.

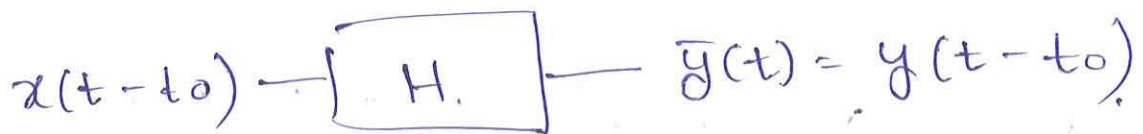
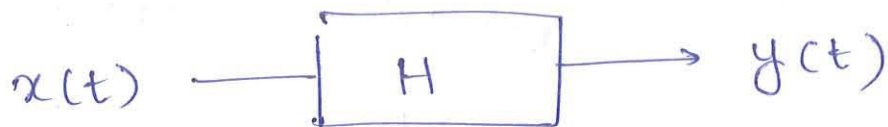


linear? Answer no! (why?)

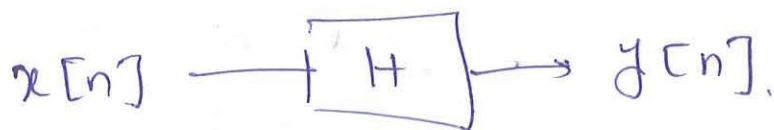
Time invariant system.

A system H is said to be time invariant if for any input $x(t)$ we get $y(t) = H\{x(t)\}$ then

for $x(t - t_0)$ (for any $t_0 \in \mathbb{R}$) we must get $y(t - t_0)$ as the output.

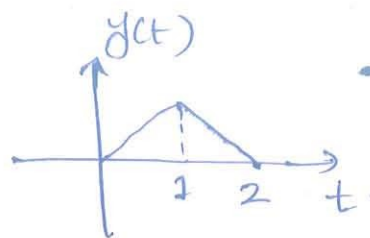
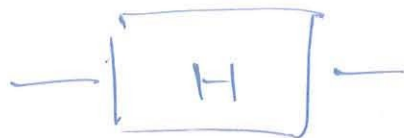
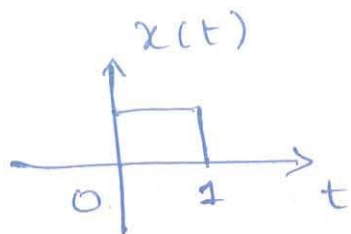


In discrete time case (also called shift invariance)

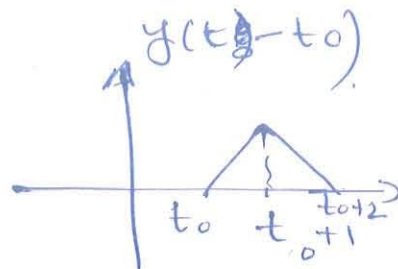
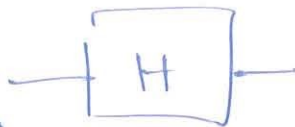
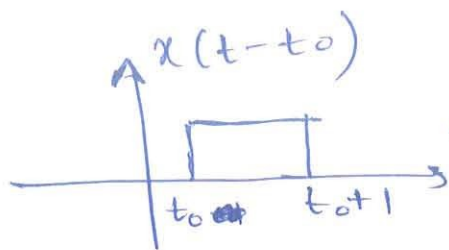


eg:

if



then



- Systems with ^{and} without memory

(i) Memoryless Systems:

Output depends upon only the input value at current instance of time.

eg: $y(t) = 2x(t)$

(ii) Memory.

Output has dependence on ^{past} input ^{future} / or both.
Value.

ex: 1) $y(t) = x(t-1) + x(t+1)$

2) $y(t) = \int_{-\infty}^t x(\tau) d\tau$

3) $y(t) = \frac{d}{dt} x(t)$

Invertibility of system.

A system is invertible if it is possible to uniquely infer input given the output signal.

In other words

Distinct inputs leads to distinct output.

Ideally we would like ~~to~~ systems to be invertible (Think of encoding - decoding)

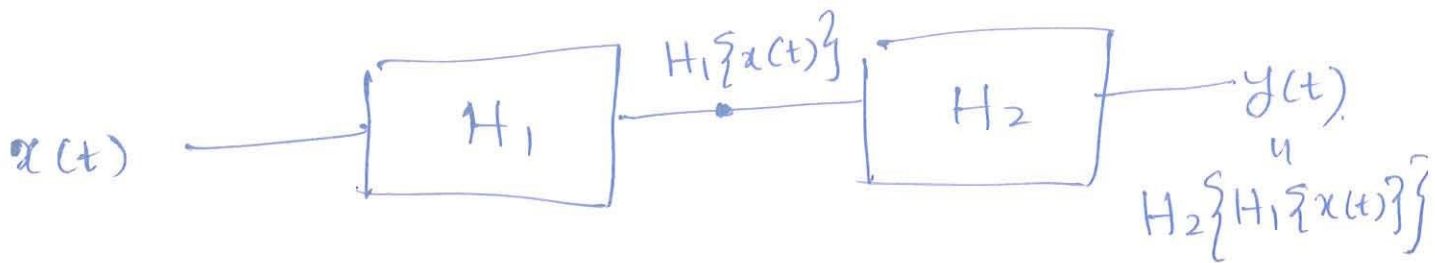
[But not for adversaries]

- Causal Systems.

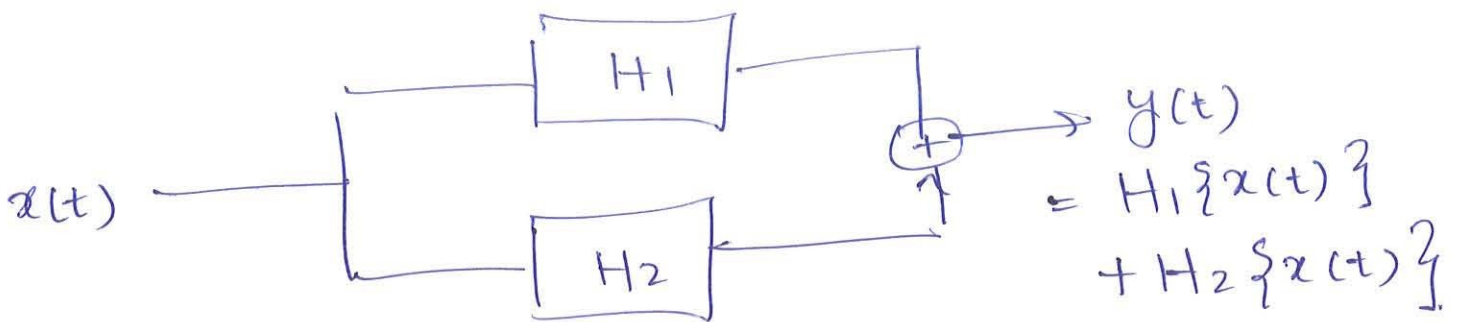
- output at any time depend only on input values at present and/or those seen in the past.
- Non-anticipating nature.
- $y(t) = x(t+1)$ (is not Causal system)
- $y[n] = x[n] + 2x[n-1]$ is causal.
- $y(t) = x(t-1)$? $y(t) = \frac{d}{dt}x(t)$?

Various interconnections of systems.

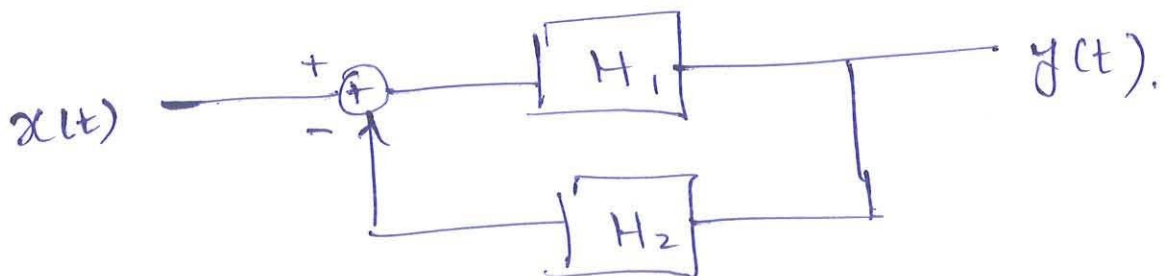
a) Cascade interconnection.



b) Parallel interconnection.



c) Feedback interconnection.



$$y(t) = H_1 \left\{ x(t) - H_2 \{ y(t) \} \right\}$$

d) Combination of a, b, c.

