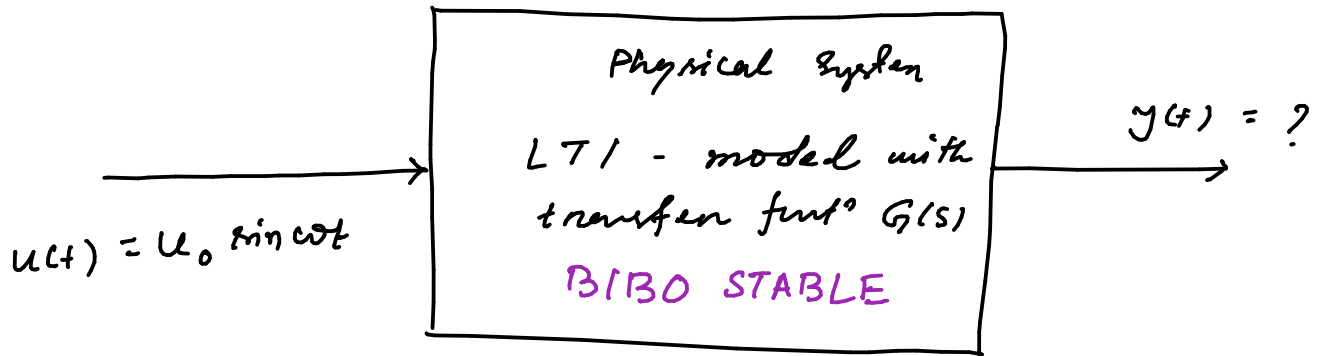


Lecture - 26

Let us assume that the transfer funⁿ $G(s) = \frac{n(s)}{d(s)}$ has distinct poles.
 (deg $n(s) < \text{deg } d(s)$)
 ↓
 poles: p_i

The output

$$Y(s) = G(s) U(s)$$

$$= G(s) \frac{u_0 \omega}{s^2 + \omega^2}$$

Using partial fraction expansion .

$$Y(s) = G(s) \frac{u_0 \omega}{s^2 + \omega^2}$$

$$= \underbrace{\frac{a}{s + j\omega} + \frac{\bar{a}}{s - j\omega}} + \frac{b_1}{s + p_1} + \frac{b_2}{s + p_2} + \dots + \frac{b_n}{s + p_n}$$

\bar{a} is complex conjugate of a

We have assumed that $G(s)$ is BIBO stable

\Downarrow

the poles p_i are in open left half of \mathbb{C} .

$$Y(s) = \frac{a}{s+j\omega} + \frac{\bar{a}}{s-j\omega} + \frac{b_1}{s+p_1} + \frac{b_2}{s+p_2} + \dots + \frac{b_n}{s+p_n}$$

$\mathcal{L}^{-1} \downarrow$

$$y(t) = a e^{-j\omega t} + \bar{a} e^{j\omega t} + \underbrace{b_1 e^{-p_1 t} + b_2 e^{-p_2 t} + \dots + b_n e^{-p_n t}}_{\downarrow}$$

At steady state ($t \rightarrow \infty$)
these terms will decay down
to 0, since p_i have -ve
real parts.

At steady state :

$$y_{ss}(t) = a e^{-j\omega t} + \bar{a} e^{j\omega t}$$

\hookrightarrow To find a & \bar{a} , let us take the
Laplace transform :

$$Y_{ss}(s) = G(s) \frac{\omega u_0}{s^2 + \omega^2} = \frac{a}{s+j\omega} + \frac{\bar{a}}{s-j\omega}$$

$$\Rightarrow G(s) \omega u_0 = a(s-j\omega) + \bar{a}(s+j\omega) \quad \text{--- } \textcircled{*}$$

For $s = -j\omega$

$$G(-j\omega) \omega u_0 = a (-j\omega - j\omega)$$

$$\Rightarrow a = - \frac{G(-j\omega) u_0}{2j}$$

For $s = j\omega$

$$G(j\omega) \omega u_0 = \bar{a} (2j\omega)$$

$$\Rightarrow \bar{a} = \frac{G(j\omega) u_0}{2j}$$

For any $\omega > 0$, $G(j\omega)$ is a complex number, which can be represented as

$$G(j\omega) \rightarrow |G(j\omega)| e^{j\varphi}$$

$$\text{When } \varphi = \angle G(j\omega) = \tan^{-1} \left[\frac{\text{Im. part of } G(j\omega)}{\text{Real part of } G(j\omega)} \right]$$

$$G(-j\omega) = |G(j\omega)| e^{-j\varphi}$$

Then

$$a = - \frac{u_0 |G(j\omega)| e^{-j\varphi}}{2j}$$

$$\bar{a} = \frac{u_0 |G(j\omega)| e^{j\varphi}}{2j}$$

$$y_{ss}(t) = a e^{-j\omega t} + \bar{a} e^{j\omega t}$$

$$= - \frac{u_0 |G(j\omega)| e^{-j\varphi}}{2j} e^{-j\omega t} + \frac{u_0 |G(j\omega)| e^{j\varphi}}{2j} e^{j\omega t}$$

$$= u_0 |G(j\omega)| \frac{e^{j(\omega t + \varphi)} - e^{-j(\omega t + \varphi)}}{2j}$$

$$= \underbrace{u_0 |G(j\omega)|}_{y_0} \sin(\omega t + \varphi)$$

$$= y_0 \sin(\omega t + \varphi)$$

• At steady state, the output response of an LTI BIBO stable system will produce

$$y_{ss}(t) = y_0 \sin(\omega t + \varphi) \quad \text{for an input}$$

$$u(t) = u_0 \sin \omega t$$

when

$$y_0 = u_0 |G(j\omega)| \quad \varphi = \angle G(j\omega)$$

We will see only changes in magnitude y_0 and phase angle ϕ .

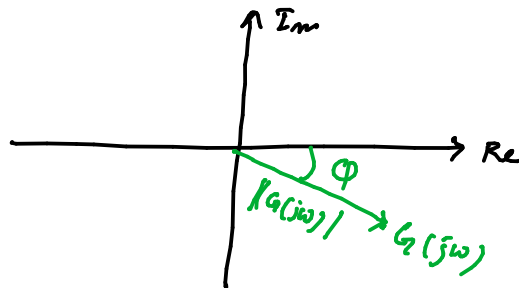
- Both $y_0 = u_0 |G(j\omega)|$ and $\phi = \angle G(j\omega)$ are functⁿ of frequency.

- By changing the frequency, we will only see changes in $|G(j\omega)|$ & $\angle G(j\omega)$

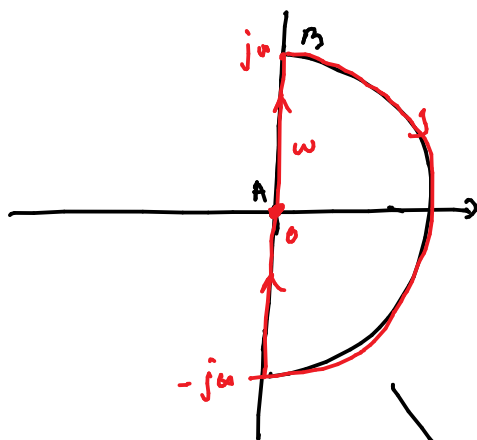


To plot $|G(j\omega)|$ & $\angle G(j\omega)$ we can

- plot in complex plane



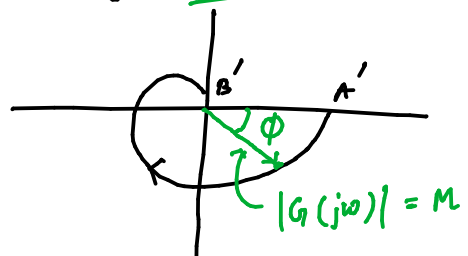
← we studied in Nyquist plot



We are only interested for frequency range

$0 \leq \omega < \infty$

mapped thru $G(j\omega)$



The magnitude & phase angles of $G(j\omega)$ are combined together & represented in complex plane. Polan plot

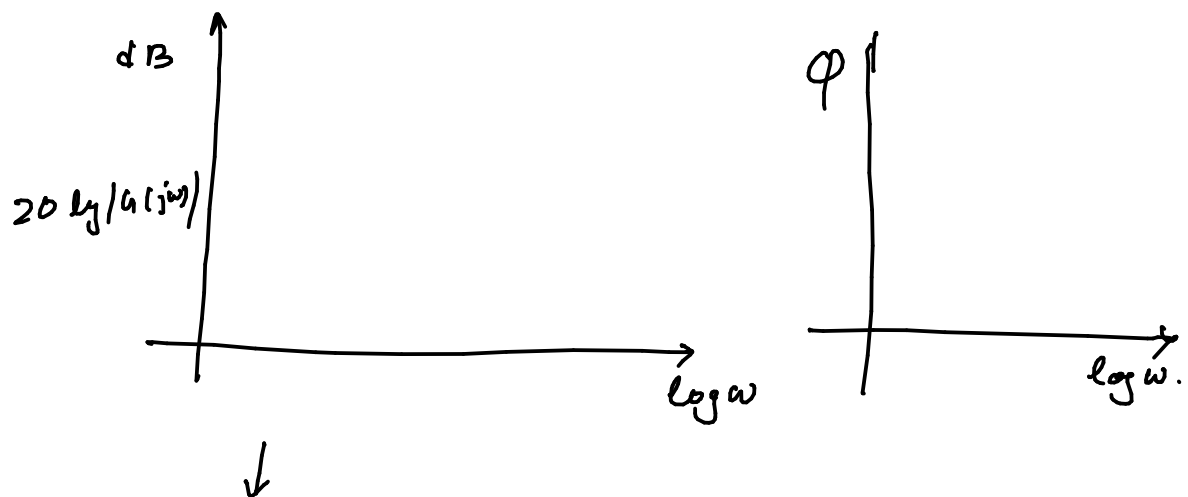
- Also the magnitude & phase can be plotted separately for varying frequency.

Usually the magnitudes are expressed in terms of dB.

$$\hookrightarrow |G(j\omega)| \rightarrow 20 \log |G(j\omega)| \text{ dB} \quad (\text{decibel})$$

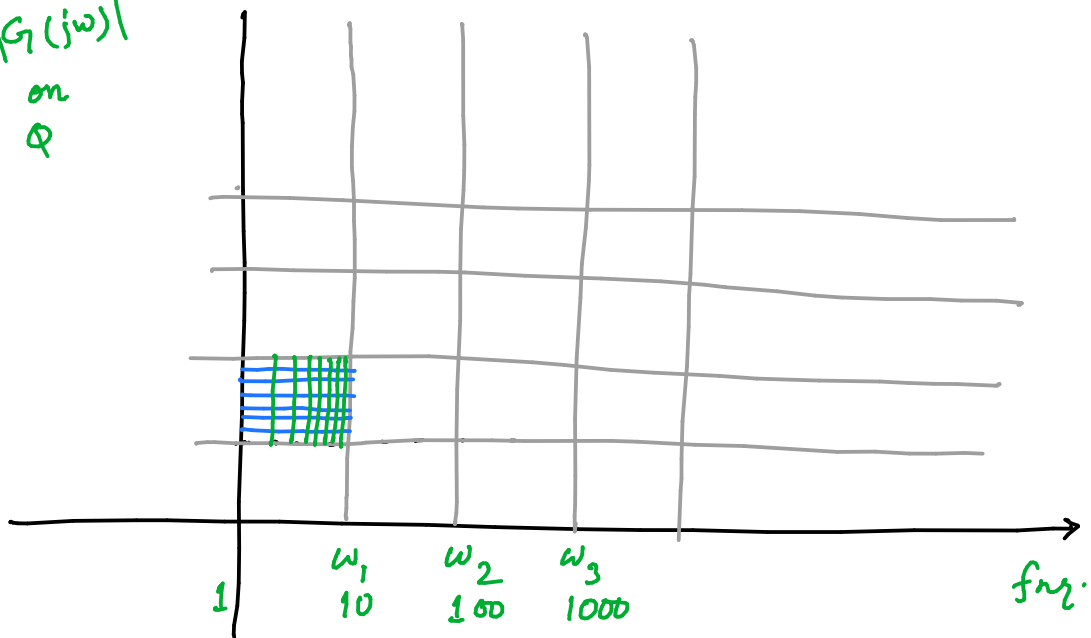
↓

- The magnitude & phase of $G(j\omega)$ will be plotted in terms of dB ~ $\log \omega$.



We use semi-log paper, where the horizontal axis is in terms of logarithmic scale & vertical axis is in linear scale.

20 log |G(jw)|
on
φ



$$\log \omega_2 - \log \omega_1$$

$$= 2 \log 10 - \log 10 = 1$$

$$\log \omega_3 - \log \omega_2$$

$$= 3 \log 10 - 2 \log 10$$

$$= 1$$

"decade" ↙

"decade"

→ Plotting of frequency response (by varying ω) plot:

Bode
Plot

- magnitude plot $20 \log |G(j\omega)| \sim \log \omega$
- phase plot $\phi = \angle G(j\omega) \sim \log \omega$

Assume that

$$G(s) = \frac{K (s+z_1)(s+z_2) \dots (s+z_m)}{s^N (s+p_1)(s+p_2) \dots (s+p_n)}$$

$$|G(j\omega)| = \frac{|K| |(s+z_1)| |(s+z_2)| \dots |(s+z_m)|}{|s^N| |(s+p_1)| |(s+p_2)| \dots |(s+p_n)|} \Bigg|_{s \rightarrow j\omega}$$

$$\lg(A/B) = \lg A - \lg B$$

$$\lg(A \cdot B) = \lg A + \lg B$$

$$20 \lg |G(j\omega)| = 20 \lg |K| + 20 \lg |j\omega + z_1| + \dots + 20 \lg |j\omega + z_m| \\ - 20 \lg |(j\omega)^N| - 20 \lg |j\omega + p_1| - \dots - 20 \lg |j\omega + p_n|$$

$$\phi = \sum_{i=1}^m \theta_{z_i} - \sum_{k=1}^{n+N} \theta_{p_k}$$

→ Basic factors that appear in $G(s)$

- Gain K
- Integral & derivative factor $\frac{1}{s}$ on s ($(j\omega)^{\pm 1}$)
- First order factors $|s + z_i|$ or $\frac{1}{|s + p_i|}$
 $(1 + j\omega\tau)^{\pm 1}$
- Quadratic terms $(1 + 2\zeta(j\omega/\omega_n) + j(\omega/\omega_n)^2)^{\pm 1}$

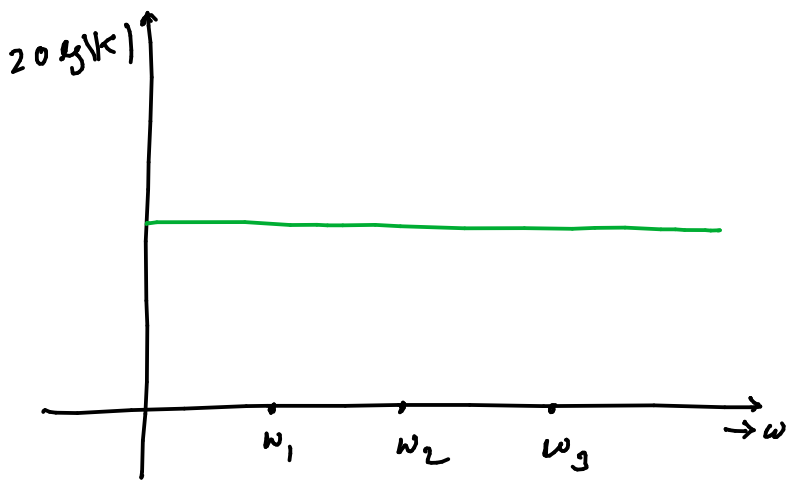
→ For gain K

Evaluate $20 \log |K|$ → at different frequencies

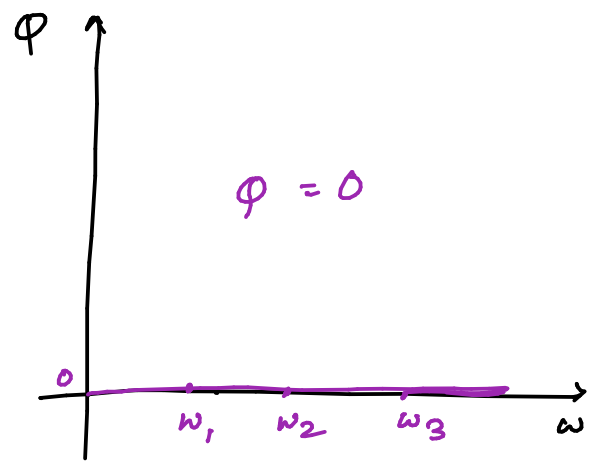
$$\omega_1 = 0.1$$

$$\omega_2 = 10$$

$$\omega_3 = 100$$



Magnitude plot



Phase plot