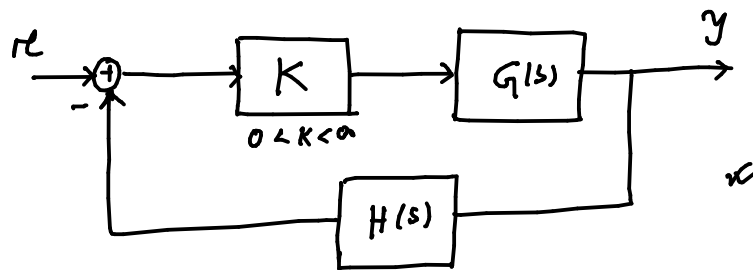


Lecture - 30

$$G(s) = \frac{n_g(s)}{d_g(s)}$$

$$H(s) = \frac{n_h(s)}{d_h(s)}$$

Closed loop T.F. $T(s) = \frac{KG(s)}{1 + KG(s)H(s)}$

$$= \frac{K n_g d_h}{d_g d_h + K n_g n_h} \checkmark$$

Closed loop characteristic polynomial

$$\sigma(s) = d_g(s)d_h(s) + K n_g(s)n_h(s)$$

roots of $\sigma(s)$ are the closed loop poles

Root locus: Locus of roots of the polynomial $\sigma(s)$ with change in K within $0 \leq K < \infty$

→ Sketching of Root locus

the path traversed by a closed loop pole in \mathbb{C} .

① The number of branches in a root locus is equal to the number of closed loop poles.

② Since the polynomial $\sigma(s)$ has real-coefficients, the closed loop poles will appear with complex conjugate.

The root locus is symmetric about real axis.

- Starting & Ending points of root-locus:

$$\underline{\underline{0 \leq K < \infty}}$$

$$\sigma(s) = d_g(s) d_h(s) + K n_g(s) n_h(s)$$

For small value of K

$$\sigma(s) = d_g(s) d_h(s) + \epsilon \quad \leftarrow \text{small number.}$$

$$\approx d_g(s) d_h(s) \quad \leftarrow \text{the } \underline{\text{poles of } G(s)H(s)}$$

For large value of K

$$\sigma(s) \approx K n_g(s) n_h(s)$$

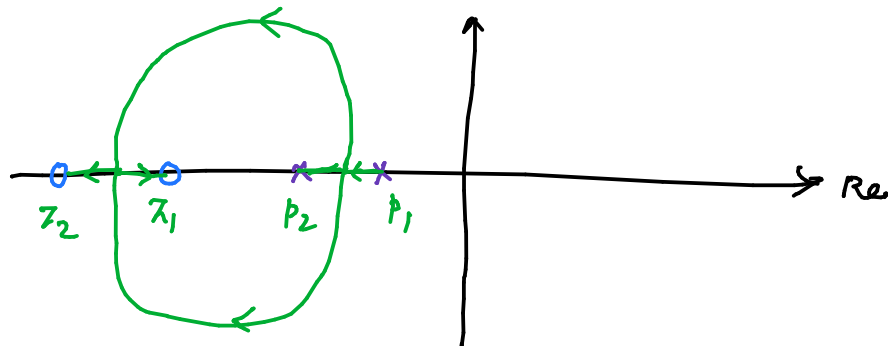
↑

zeros of G(s)H(s)

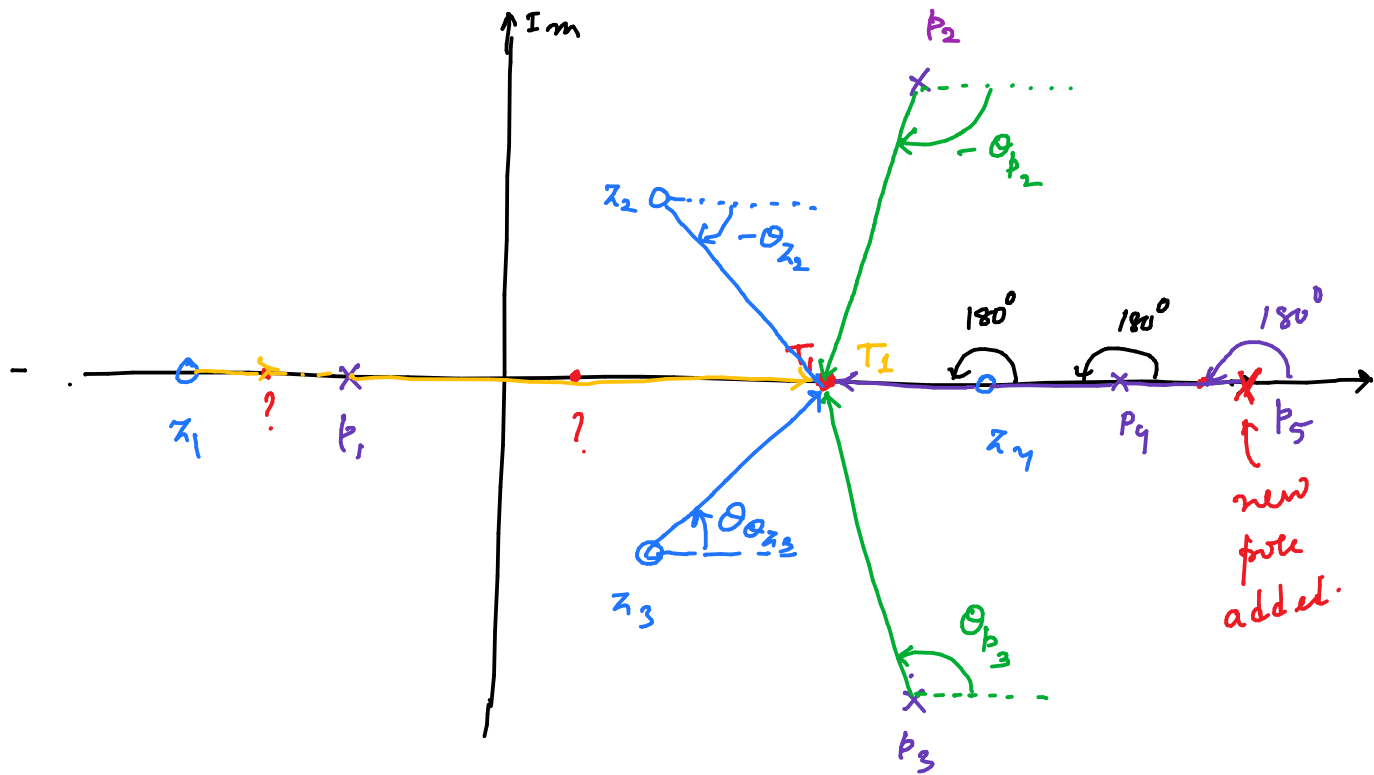
- The root locus begins at the finite & infinite poles of $G(s)H(s)$ & ends at the finite & infinite zeros of $G(s)H(s)$.

$$G(s)H(s) = \frac{(s+3)(s+4)}{(s+1)(s+2)}$$

← It has finite number of poles & zeros.



→ Real axis segment



- A complex no. s is on the root locus

if

$$\begin{cases} |K G(s) H(s)| = 1 \\ \angle K G(s) H(s) = (2L+1)180^\circ \quad L=0, \pm 1, \pm 2 \dots \end{cases}$$

θ

$$\theta = \theta_{z_1} + \theta_{z_2} + \theta_{z_3} + \theta_{z_4} - \theta_{p_1} - \theta_{p_2} - \theta_{p_3} - \theta_{p_4}$$

→ θ_{p_2} & θ_{p_3} are equal angle but opposite sign. So the net angular contribution to form θ due to θ_{p_2} & $\theta_{p_3} = 0^\circ$

→ θ_{z_2} & θ_{z_3} are equal angles but they have opposite sign. So net angular contribution to form θ due to θ_{z_2} & $\theta_{z_3} = 0^\circ$

- At each point on the real axis of \mathbb{C} , the angular contribution for Θ due to a pair of complex conjugate poles or zeros is 0° .

→ The angular contribution at T_1 due to p_1 & z_1 is 0° .

- The angular contribution of poles & zeros of $G(s)H(s)$, which are on real axis and left to the test point, is 0° .

- The only contribution to form Θ at test point T_1 is due to the poles & zeros of $G(s)H(s)$, which are on the real axis of \mathbb{C} and to the right of test point T_1 .

→ The angular contribution at T_1 is then only due to z_4 & p_4

$$\Theta = \Theta_{z_4} - \Theta_{p_4} = 180^\circ - 180^\circ = 0^\circ$$

⇓

T_1 is not on the root locus

(we have even number of poles & zeros to the right of T_1)

By adding one extra pole to the right of T_1

(We have now odd number of poles & zeros
+ the right of T_1)

↓

$$\begin{aligned}\theta &= \theta_{z_4} - \theta_{p_4} - \theta_{p_5} \\ &= 180^\circ - 180^\circ - 180^\circ = -180^\circ\end{aligned}$$

↓

T_1 is on the root locus.

→ On the real axis, for $K > 0$ the root locus exists to the left of an odd number of real axis finite poles and/or zeros of $G(s)H(s)$.

→ Behavior at infinity

$$G(s) = \frac{n(s)}{d(s)}$$

For any value of s s.t. $G(s) \rightarrow \infty$

then s is a pole of $G(s)$

For any value of s s.t. $G(s) \rightarrow 0$,

then s is a zero of $G(s)$.

- A pole of $G(s)$ is at ∞ if

$$G(s) \rightarrow \infty \text{ as } s \rightarrow \infty$$

Ex : $G(s) = s$

- A zero of $G(s)$ is at ∞ if

$$G(s) \rightarrow 0 \text{ as } s \rightarrow \infty$$

Ex. $G(s) = \frac{1}{s}$

$$G(s) = \frac{(s+3)}{(s+1)(s+4)(s+5)}$$

Finite poles : $-1, -4, -5$

Finite zeros : -3

$$G(s) \approx \frac{1}{s \cdot s} \text{ For } s \rightarrow \infty$$

It has 2 zeros at ∞ .

It has no poles at ∞

- For every transfer function $G(s)$, there are equal number of poles & zeros if we include the poles & zero at finite as well as infinite position.

$$G(s) = \frac{(s+2)(s+3)}{(s+2)} \begin{cases} \text{It has No zeros at } \infty \\ \text{It has One pole at } \infty \end{cases}$$

By changing K , the real roots of a quadratic polynomial, become complex conjugate

$$K G(s) = \frac{K}{(s+1)(s+4)}$$

↓

Closed loop ch. polynomial

$$\alpha(s) = s^2 + 5s + 4 + K$$

$$s = -1, -4 \quad \text{for } K=0$$

$$s = -3.62, -1.38 \quad K=1$$

$$s = -3, -2 \quad K=2$$

$$s = -2.5 \pm j0.86 \quad K=3$$

$$s = -2.5 \pm j2.78 \quad K=10$$

