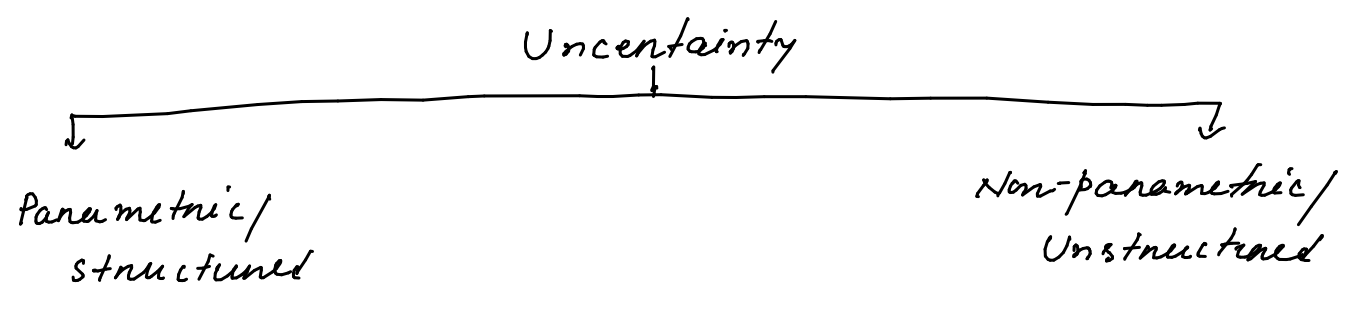


# Lecture-16

## Unstructured Uncertainty:

$\Omega_{pr}^s \rightarrow$  proper & stable t-fs



•  $\delta \in \mathbb{R}$  is an uncertain parameter

•  $\Delta(s) \in \Omega_{pr}^s$  is an uncertain parameter

• Ex:  $P(s, \delta) = \frac{1}{s^2 + 2\delta_1 s + 3\delta_2}$

$\delta_1 \in \mathbb{R}$   
 $\delta_2 \in \mathbb{R}$

Additive

Multiplicative

$G_\Delta(s) = G_0(s) + \Delta(s)$

$G_\Delta(s) = (1 + \Delta(s)) G_0(s)$

$G_0(s)$  is the nominal plant.



For stability study & control design we need to quantify the uncertainty in some sense

• Quantification:  $\delta_i \in [\delta_i^-, \delta_i^+]$

•  $\Delta(s) \in \Omega_{pr}^s$   
 $\|\Delta(s)\|_{\infty} \leq \gamma$   
where  $\gamma > 0$ .

$$P_a := \left\{ G_\Delta(s) := G_0 + \Delta(s) \mid \Delta(s) \in \Sigma_{pr}^s \right. \\ \left. \text{and } \|\Delta(s)\|_{\mathcal{H}_\infty} \leq r \right\}$$

$\downarrow$   
 $\Delta(s)$   
 $\downarrow$   
 $W_1(s) \Delta(s) W_2(s)$

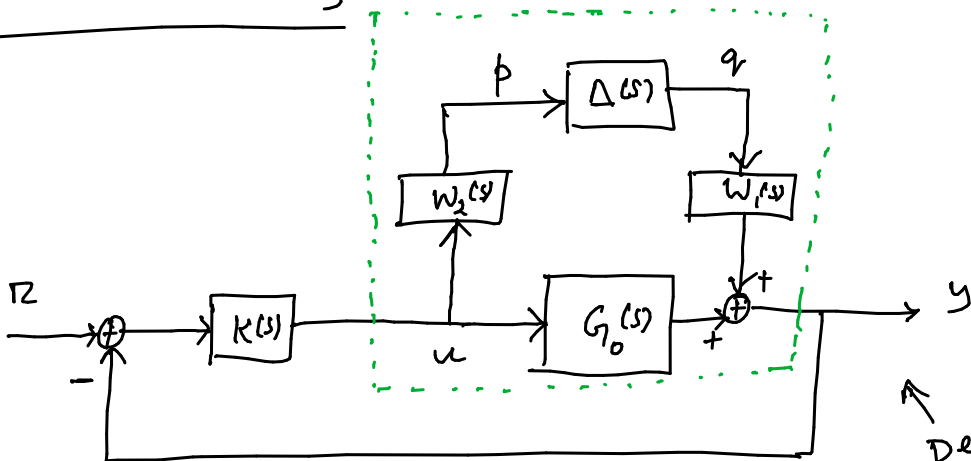
weightij function ( $W_1(s)$  and  $W_2(s)$ )  
 both belongs to  $\Sigma_{pr}^s$

$$P_a := \left\{ G_\Delta(s) = G_0(s) + W_1(s) \Delta(s) W_2(s) \mid \|\Delta(s)\|_{\mathcal{H}_\infty} \leq r \right\}$$

$$P_m := \left\{ G_\Delta(s) = (I + W_1(s) \Delta(s) W_2(s)) G_0(s) \mid \|\Delta(s)\|_{\mathcal{H}_\infty} \leq r \right\}$$

$\rightarrow$   
 multiplication  
 uncertain  
 plant family

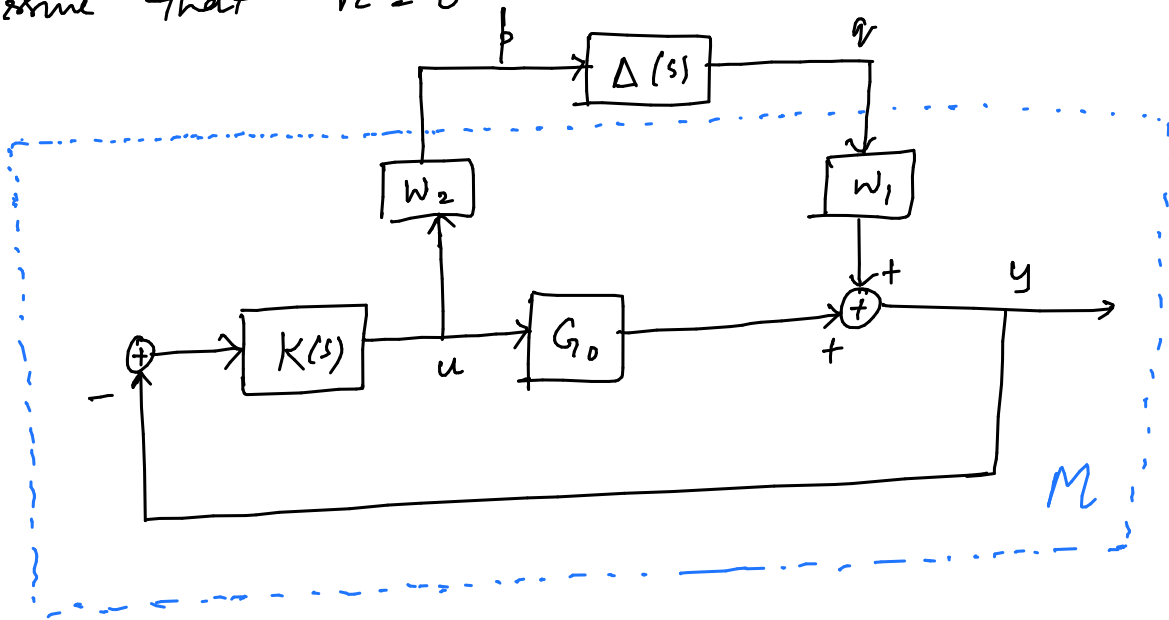
Additive Uncertainty



$\uparrow$   
 Deals with  
 Additive uncertainty

$$\begin{aligned}
 y &= G_0 u + W_1 q \\
 &= G_0 u + W_1 \Delta W_2 u = (G_0 + W_1 \Delta W_2) u \\
 y &= G_\Delta(s) u
 \end{aligned}$$

Assume that  $r = 0$



For the block  $M$ , the o/p  $\rightarrow p$   
 & i/p  $\rightarrow q$

We bring a relationship between  $p$  &  $q$

$$y = G_0 u + W_1 q$$

$$= -G_0 K y + W_1 q$$

$$\Rightarrow (I + G_0 K) y = W_1 q$$

$$\Rightarrow y = (I + G_0 K)^{-1} W_1 q$$

$$p = W_2 u$$

$$= -W_2 K y$$

$$= -W_2 K (I + G_0 K)^{-1} W_1 q$$

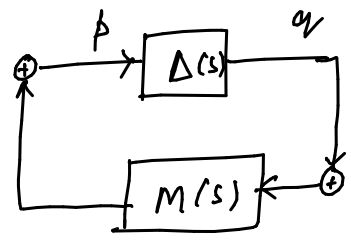
$$\left. \begin{aligned} p &= -W_2 K (I + G_0 K)^{-1} W_1 q \\ &= -W_2 K S_0 W_1 q \end{aligned} \right\}$$

$$S_0 = (I + G_0 K)^{-1}$$

sensitivity fun<sup>n</sup>?

$$\text{So } M = -W_2 K S_0 W_1$$

The additive feedback interconnect can be represented as

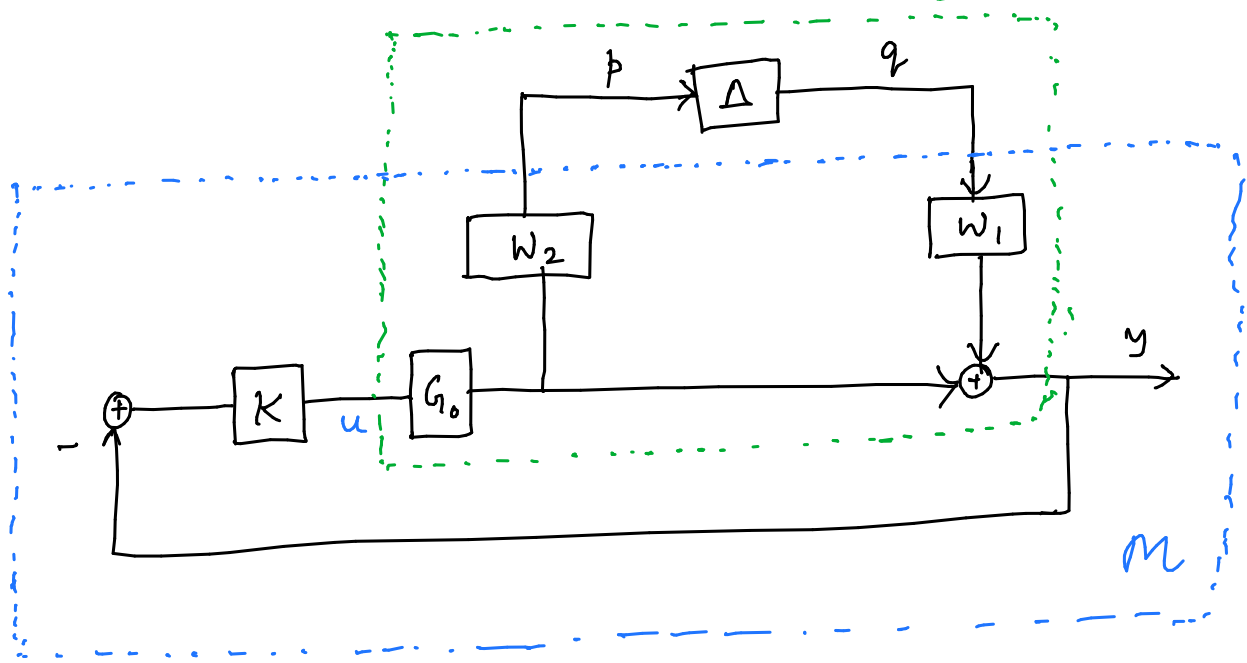


→ (M-Δ) structure

When  $M = -W_2 K S_0 W_1$

Multiplicative Uncertainty:

*for uncertainty representation*



$$\begin{aligned}
 y &= G_0 u + W_1 q \\
 &= G_0 u + W_1 \Delta W_2 G_0 u = \underbrace{(I + W_1 \Delta W_2)}_{G_\Delta(s)} G_0 u
 \end{aligned}$$

Multiplicative uncertainty

↓ Construct the M-Δ structure

$$y = G_0 u + W_1 z$$

$$p = W_2 G_0 u$$

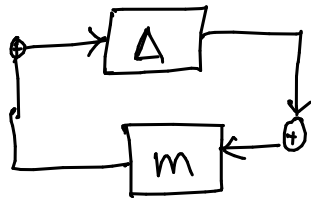
$$\Rightarrow y = (I + G_0 K)^{-1} W_1 z$$

$$= -W_2 G_0 K y$$

$$= -W_2 G_0 K (I + G_0 K)^{-1} W_1 z$$

$$\boxed{p = -W_2 T_0 W_1 z} \rightarrow \boxed{M = -W_2 T_0 W_1}$$

Where  $T_0 = G_0 K (I + G_0 K)^{-1}$  ← Complementary sensitivity function



→ Stability Study of  $M-\Delta$  structure:

↳ we use Small gain theorem for the robust stability analysis

### Small Gain Theorem

Let us assume that  $M(s) \in \Sigma_{pr}^s$ .

Then the feedback interconnection is well-posed and internally stable for all  $\Delta(s) \in \Sigma_{pr}^s(\mathbb{R}Fluo)$

with  $\|\Delta(s)\|_{Fluo} < \frac{1}{\gamma}$  if and only if  $\|M(s)\|_{Fluo} \leq \gamma$ .

(proof → see Zhou, Doyle book)

Based on Small Gain Theorem:

(2)

→ Result: Let  $K(s)$  be a stabilizing controller for the nominal plant  $G_0(s)$ . Let

$$G_A(s) = G_0(s) + W_1(s)\Delta(s)W_2(s) \quad (\text{additive uncertainty})$$

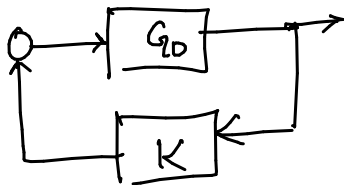
where  $\Delta(s) \in \Omega_{\text{pc}}^s$  with  $\|\Delta(s)\|_{\mathcal{H}_\infty} < \frac{1}{\gamma_a}$ .

Then the feedback interconnection (M- $\Delta$  st)

is robustly stable for all  $\|\Delta(s)\|_{\mathcal{H}_\infty} < \frac{1}{\gamma_a}$

if and only if  $\|W_2 K S_0 W_1\|_{\mathcal{H}_\infty} \leq \gamma_a$ .

Proof



Since  $K$  is a stabilizing controller for  $G_0$ , we

can say that

$$S_0 = \frac{1}{1 + G_0 K} \quad \text{is stable \& proper}$$

(For SISO case)  
 Recall that for internal stability all of the 4 t.f. are stable.

$$S_0 \quad K S_0 = \frac{K}{1 + G_0 K} \quad \text{is also stable \& proper}$$

$W_2 K S_0 W_1 \leftarrow$  is also stable \& proper

Since  $W_2$  \&  $W_1$  are stable \& proper

$$\|W_2 K S_0 W_1\|_{\mathcal{H}_\infty}$$

can now be defined. Then the result directly follows from SGT.

The design parameter is  $K(s)$ .

(7)

→ Result:

Let  $K(s)$  be a stabilizing controller for the nominal plant  $G_0(s)$ . Let

$$G_\Delta(s) = (I + W_1 \Delta W_2) G_0 \quad (\text{multiplication uncertainty}).$$

When  $\Delta(s) \in \Sigma_{pr}^s$ . The the feedback interconnection is well-posed and internally stable (Robustly stable) for all

$$\|\Delta(s)\|_{\mathcal{H}_\infty} < \frac{1}{r} \quad \text{if and only if} \quad \|W_2 T_0 W_1\|_{\mathcal{H}_\infty} \leq r.$$

For Robust stability

- Additive uncertainty  $\rightarrow \|W_2 K S_0 W_1\|_{\mathcal{H}_\infty} \leq r$
- Multiplication uncertainty  $\rightarrow \|W_2 T_0 W_1\|_{\mathcal{H}_\infty} \leq r$ .

provided  $\|\Delta\|_{\mathcal{H}_\infty} < \frac{1}{r}$



→ A common platform to study all kind of uncertainties:



Linear Fractional Transformation

## → Linear Fractional Transformation:

②

Let us consider a complex matrix  $M$  which is partitioned as follows:

$$M = \begin{bmatrix} M_{11} & | & M_{12} \\ \hline M_{21} & | & M_{22} \end{bmatrix}$$

Let  $\Delta_L$  &  $\Delta_U$  be other two complex matrices.

Then define

- Lower LFT:

$$F_L(M, \Delta_L) := M_{11} + M_{12} \Delta_L (I - M_{22} \Delta_L)^{-1} M_{21}$$

provided  $(I - M_{22} \Delta_L)^{-1}$  exists.

- Upper LFT

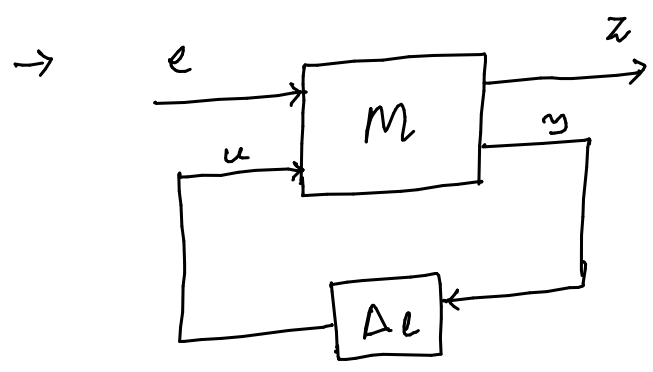
$$F_U(M, \Delta_U) := M_{22} + M_{21} \Delta_U (I - M_{11} \Delta_U)^{-1} M_{12}$$

provided  $(I - M_{11} \Delta_U)^{-1}$  exists.

- The LFTs are called "well-defined" iff

$$(I - M_{11} \Delta_U)^{-1} \text{ \& \ } (I - M_{22} \Delta_L)^{-1} \text{ exist.}$$





$$M = \begin{bmatrix} m_{11} & m_{12} \\ m_{21} & m_{22} \end{bmatrix}$$

$$\begin{bmatrix} z \\ y \end{bmatrix} = \begin{bmatrix} m_{11} & m_{12} \\ m_{21} & m_{22} \end{bmatrix} \begin{bmatrix} e \\ u \end{bmatrix}$$

Bring a relationship bet<sup>n</sup> z & e

$$u = \Delta_L y$$

$$z = m_{11}e + m_{12}u$$

$$y = m_{21}e + m_{22}u$$

$$= m_{11}e + m_{12}\Delta_L y$$

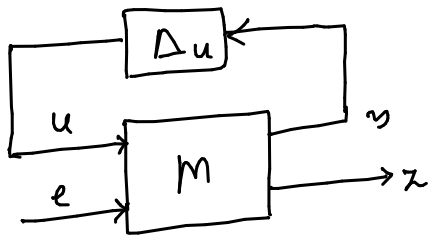
$$= m_{21}e + m_{22}\Delta_L y$$

$$\Rightarrow y = (\mathbf{I} - m_{22}\Delta_L)^{-1} m_{21}e$$

$$z = m_{11}e + m_{12}\Delta_L (\mathbf{I} - m_{22}\Delta_L)^{-1} m_{21}e$$

$$= \underbrace{(m_{11} + m_{12}\Delta_L (\mathbf{I} - m_{22}\Delta_L)^{-1} m_{21})}_F e$$

$$F_L(M, \Delta_L)$$



$$M = \begin{bmatrix} M_{11} & M_{12} \\ M_{21} & M_{22} \end{bmatrix}$$

$$\begin{bmatrix} y \\ z \end{bmatrix} = \begin{bmatrix} M_{11} & M_{12} \\ M_{21} & M_{22} \end{bmatrix} \begin{bmatrix} u \\ e \end{bmatrix}, \quad u = \Delta_u y$$

$$y = M_{11}u + M_{12}e$$

$$z = M_{21}u + M_{22}e$$

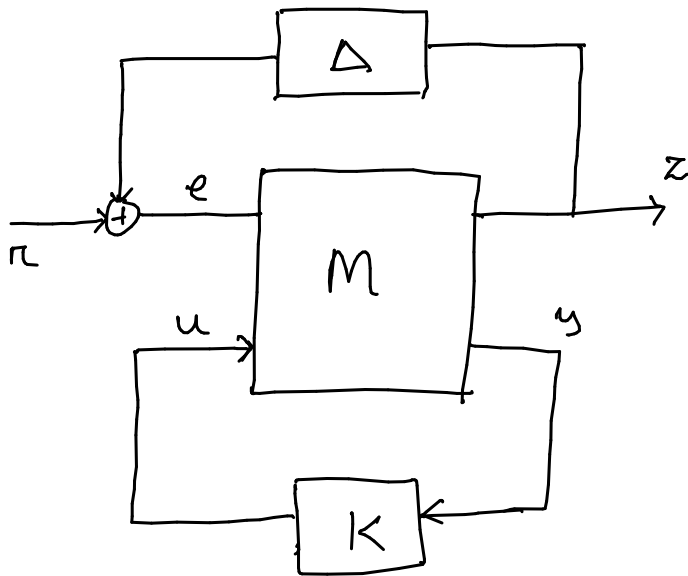
$$= M_{21}\Delta_u y + M_{22}e$$

$$\Rightarrow y = (\mathbf{I} - M_{11}\Delta_u)^{-1} M_{12}e$$

$$= \underbrace{\left( M_{22} + M_{21}\Delta_u(\mathbf{I} - M_{11}\Delta_u)^{-1} M_{12} \right)}_{F_u(M, \Delta_u)} e$$

$$z = F_u(M, \Delta_u)e$$

→ A common configuration for Robust stability:-



$$M = \begin{bmatrix} M_{11} & M_{12} \\ M_{21} & M_{22} \end{bmatrix}$$

Assume  $r=0$ ,  $z=0$   $K=0$

When  $K=0$ , we are dealing with upper LFT

$$\begin{aligned} \text{So } z &= M_{11}e + M_{12}u \\ &= M_{11}\Delta z + M_{12}u \\ \Rightarrow z &= (\mathbf{I} - M_{11}\Delta)^{-1} M_{12}u \end{aligned}$$

$$\begin{aligned} y &= M_{21}e + M_{22}u \\ &= M_{21}\Delta z + M_{22}u \\ &= M_{21}\Delta (\mathbf{I} - M_{11}\Delta)^{-1} M_{12}u + M_{22}u \\ &= \underbrace{\left[ M_{22} + M_{21}\Delta (\mathbf{I} - M_{11}\Delta)^{-1} M_{12} \right]}_{F_u(M, \Delta)} u \end{aligned}$$

$y \rightarrow u$  relation was for representing  
the type of uncertainty we deal.

$$y = G_{\Delta}(s)u$$

when  $G_{\Delta}(s) = M_{22} + M_{21}\Delta (\mathbf{I} - M_{11}\Delta)^{-1} M_{12}$

Let us now choose  $M_{11} = 0$   $M_{12} = W_2$   $M_{21} = W_1$

$$\text{and } M_{22} = G_0$$

$$M = \begin{bmatrix} 0 & W_2 \\ W_1 & G_0 \end{bmatrix} \rightarrow G_{\Delta}(s) = G_0 + W_2 \Delta W_2$$

↑  
additive uncertainty

let us use

$$M = \begin{bmatrix} 0 & : & W_2 G_0 \\ \dots & : & \dots \\ W_1 & : & G_0 \end{bmatrix}$$

Then  $G_A(s) \approx G_0 + W_1 \Delta W_2 G_0$

$$= (\mathbb{I} + W_1 \Delta W_2) G_0$$

↑  
Multiplicative uncertainty